

# Voltage- and current-activated metal–insulator transition in VO<sub>2</sub>-based electrical switches: a lifetime operation analysis

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## Abstract

Vanadium dioxide is an intensively studied material that undergoes a temperature-induced metal–insulator phase transition accompanied by a large change in electrical resistivity. Electrical switches based on this material show promising properties in terms of speed and broadband operation. The exploration of the failure behavior and reliability of such devices is very important in view of their integration in practical electronic circuits. We performed systematic lifetime investigations of two-terminal switches based on the electrical activation of the metal–insulator transition in VO<sub>2</sub> thin films. The devices were integrated in coplanar microwave waveguides (CPWs) in series configuration. We detected the evolution of a 10 GHz microwave signal transmitted through the CPW, modulated by the activation of the VO<sub>2</sub> switches in both voltage- and current-controlled modes. We demonstrated enhanced lifetime operation of current-controlled VO<sub>2</sub>-based switching (more than 260 million cycles without failure) compared with the voltage-activated mode (breakdown at around 16 million activation cycles). The evolution of the electrical self-oscillations of a VO<sub>2</sub>-based switch induced in the current-operated mode is a subtle indicator of the material properties modification and can be used to monitor its behavior under various external stresses in sensor applications.

Keywords: vanadium dioxide, electrical switching, metal–insulator transition, lifetime operation

## 1. Introduction

Vanadium dioxide (VO<sub>2</sub>) exhibits a reversible temperature-driven metal–insulator transition (MIT), which markedly changes its electronic and optical properties [1, 2]. Below the transition temperature of ~68 °C, VO<sub>2</sub> behaves as an insulator or semiconductor with a monoclinic crystal structure and a band gap of about 1 eV, whereas for temperatures higher

than 68 °C, it transforms abruptly to a metallic state with a tetragonal rutile structure. In VO<sub>2</sub> thin films, this transition can be triggered by thermal [1–3], electrical (charge injection or Joule heating) [2–4] or optical excitation (photon excitation) [5, 6], and even by external pressure or strain [7]. The MIT induces extremely fast and abrupt changes in the electronic and optical properties of the material: the electrical resistivity increases by 3 to 5 orders

of magnitude (depending on the crystalline quality of the deposited films [8], stoichiometry and doping [9]) while the optical reflectivity markedly decreases [5, 6]. An activation time as short as 100 fs has been reported for the optically driven MIT transition [5, 6], and the electronically induced transition occurs within nanoseconds [2–4, 10]. The physical mechanisms underlying the MIT in VO<sub>2</sub> are not fully elucidated and it is still unclear whether the transition is driven by the crystalline phase transition (from monoclinic to the tetragonal phase) or by electron–electron correlations (pure electronic Mott transition), although recent reports apparently favor the second mechanism [11]. The remarkable properties and broadband operation of the MIT in VO<sub>2</sub> have received ever-increasing attention from the scientific community during the last few years and have made the material an interesting candidate for fast switching with feasible applications in domains spanning over the entire electromagnetic spectrum. These applications include low-frequency two- and three-terminal electrical switches [2–4], RF-microwave switches, tunable filters [10, 12] and power limiters [13], THz metamaterial devices [14, 15] and nanoresonators [16] and optical [5, 6] components.

For the realization of practical devices, the electrical activation of VO<sub>2</sub> (in two- or three-terminal configurations) is favored over the thermal one as it offers faster activation times and easier implementation. However, the literature on the lifetime and reliability of such components is scarce. An early study by Guzman *et al* [17] on the electrical characteristics of VO<sub>2</sub> thin films obtained by the sol–gel method showed that their MIT switching properties were unaffected after 108 Joule-heating activation cycles. On the other hand, Ilinski *et al* [18] observed a rapid degradation and suppression of the MIT hysteresis loops in the electrical conductivity and optical reflectivity of amorphous VO<sub>2</sub> thin films after only a few thermal cycles across the phase transition region. More recently, Ko and Ramanathan [19] have shown that the MIT characteristics of high-quality polycrystalline VO<sub>2</sub> thin films were largely unaffected after 102 thermal cycles over the phase transition. It was suggested [19] that during thermocycling, the polycrystalline structure of VO<sub>2</sub> thin films prevents microcrack formation and hinders oxygen diffusion from VO<sub>2</sub> clusters to the nearby low-oxygen regions, as observed in the amorphous films.

Studies on the effect of recurrent electrical activation of VO<sub>2</sub> switches on their properties have not yet been reported. Therefore, in this paper, we present a methodical investigation on the reliability and lifetime of electrical switching based on the MIT in VO<sub>2</sub> thin films.

## 2. Experimental details

We fabricated two-terminal devices using VO<sub>2</sub> thin films obtained by pulsed laser deposition (PLD), which were further integrated in coplanar microwave waveguides (CPWs) [10]. We recorded the evolution of a continuous microwave (MW) signal at 10 GHz traveling through a microwave guide during the sequential activation of a VO<sub>2</sub> device. Using this detection

method, we aimed to avoid interference and feedback between the detected signal and the actuation of the VO<sub>2</sub>-based switches.

VO<sub>2</sub> thin films were deposited using reactive PLD from a high-purity (99.95%) vanadium target under an oxygen atmosphere on Al<sub>2</sub>O<sub>3</sub> *c*-type substrates. The experimental conditions are detailed elsewhere [10, 12]. The obtained 200-nm-thick VO<sub>2</sub> films were crystalline and showed a change in resistivity of about 5 orders of magnitude during a thermally induced MIT [12]. The two-terminal VO<sub>2</sub> switches were fabricated in a clean room environment [10, 12]: VO<sub>2</sub> rectangular patterns with different dimensions were defined lithographically, and gold electrodes were deposited and patterned on the VO<sub>2</sub> patterns. The electrode width was 20  $\mu$ m and the spacing was varied between 5 and 50  $\mu$ m. Current–voltage (*I*–*V*) characteristics of the two-terminal devices were recorded using a Keithley 2612A source meter in voltage or current mode.

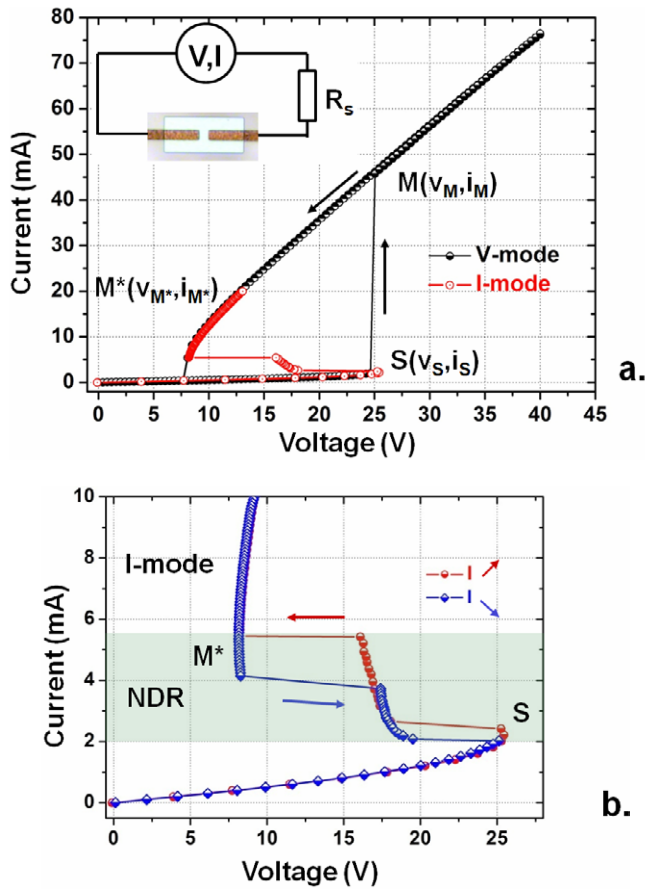
## 3. Results and discussion

### 3.1. Characterization of VO<sub>2</sub>-based two-terminal devices

Figure 1(a) shows the *I*–*V* curves (V-mode and I-mode) of a typical VO<sub>2</sub> switch of 20  $\mu$ m length and 20  $\mu$ m width. The switch was integrated in the simple measurement setup depicted in the inset of figure 1(a), in series with a 1 k $\Omega$  resistance (*R*<sub>s</sub>) and the voltage or current source. Both *I*–*V* characteristics are highly nonlinear. In the V-mode operation, when the voltage is raised to the threshold value *V*<sub>s</sub>, the VO<sub>2</sub> transforms abruptly from the highly resistive insulator state (point S on the *I*–*V* curve) to the low-resistive metallic state (point M). Consequently, the current jumps from  $\sim$ 2.2 to 46 mA. By further increasing the applied voltage beyond point M, the V-mode *I*–*V* trace follows an ohmic linear law. When decreasing the voltage, the V-mode trace shows a large hysteresis loop ( $\Delta V = 17.5$  V) until reaching point M\* where VO<sub>2</sub> transforms back to a semiconductor. The width of the hysteresis indicates an MIT transition mediated by Joule heating [3, 4, 12]. The I-mode trace is expanded in figure 1(b) with the red arrow (blue) curves corresponding to the increasing (decreasing) current between 0 and 10 mA. The clear S-type shape reveals a region of negative differential resistance (NDR) between the instability points S and M\*. It was suggested that the onset of the NDR corresponds to the MIT in a percolative manner, with the coexistence of the semiconducting and metallic domains in VO<sub>2</sub> [3, 4, 20]. Figure 1(b) reveals that the I-mode *I*–*V* curve has a narrower hysteresis (which even disappears for devices with lengths shorter than 10  $\mu$ m), indicating that, in this case, the MIT is mainly triggered by charge injection with less heat generation.

### 3.2. Coplanar microwave waveguides integrating VO<sub>2</sub> switches

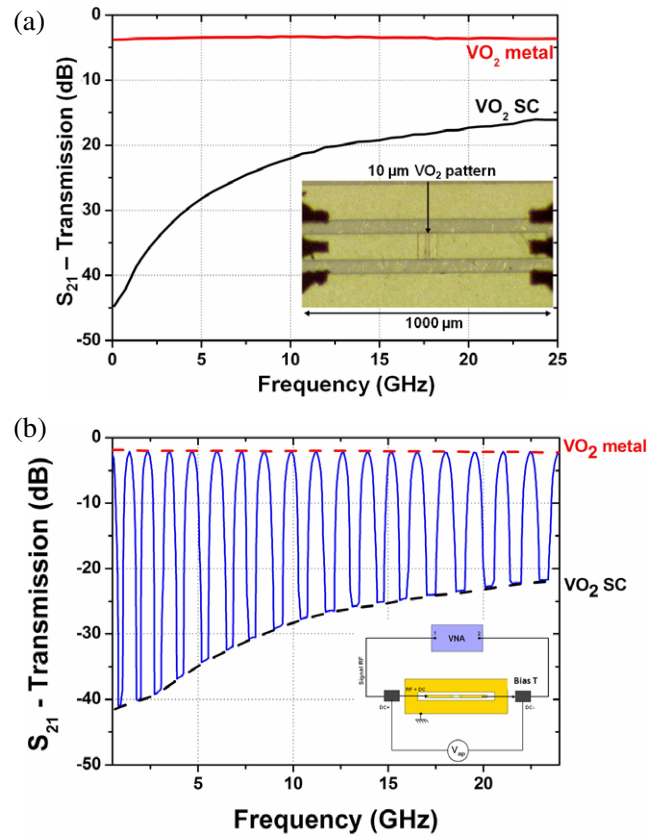
We integrated the VO<sub>2</sub> electrical switches with the *I*–*V* characteristics shown in figure 1 to coplanar microwave waveguides in a series configuration, with a central signal



**Figure 1.** (a)  $I$ - $V$  curves (V-mode and I-modes) of a two-terminal  $\text{VO}_2$  switch (20- $\mu\text{m}$  length and 20- $\mu\text{m}$  width); the inset shows the  $\text{VO}_2$  switch integrated in series with a resistance ( $R_s$ ) and the voltage or current source. (b) Enlarged view of the I-mode hysteresis indicating the boundaries of the NDR region.

line interrupted by a 10- $\mu\text{m}$ -long  $\text{VO}_2$  pattern surrounded by two ground lines and overall dimensions adapted to the 50  $\Omega$  load (inset in figure 2(a)). In-depth details of the design, fabrication and microwave performance of such devices are given elsewhere [10, 12]. The switching principle and the performance of the obtained device are presented in figure 2(a), which shows the microwave broadband transmission curves, expressed as the transmission parameter  $S_{21}$ , versus the frequency. When the  $\text{VO}_2$  is in the semiconducting state (SC), the  $\text{VO}_2$  line is highly resistive and the switch is in the off state: the microwave signal cannot travel through the device; it is attenuated by more than 15 dB between 100 MHz and 25 GHz (black curve ‘ $\text{VO}_2$  SC’ in figure 2(a)). For a threshold bias voltage of 20 V applied across the  $\text{VO}_2$  pattern,  $\text{VO}_2$  becomes metallic and the switch is in the on state: the microwave signal is transmitted through the CPW switch with insertion losses of about 3 dB between 100 MHz and 25 GHz (red curve ‘ $\text{VO}_2$  metal’ in figure 2(a)).

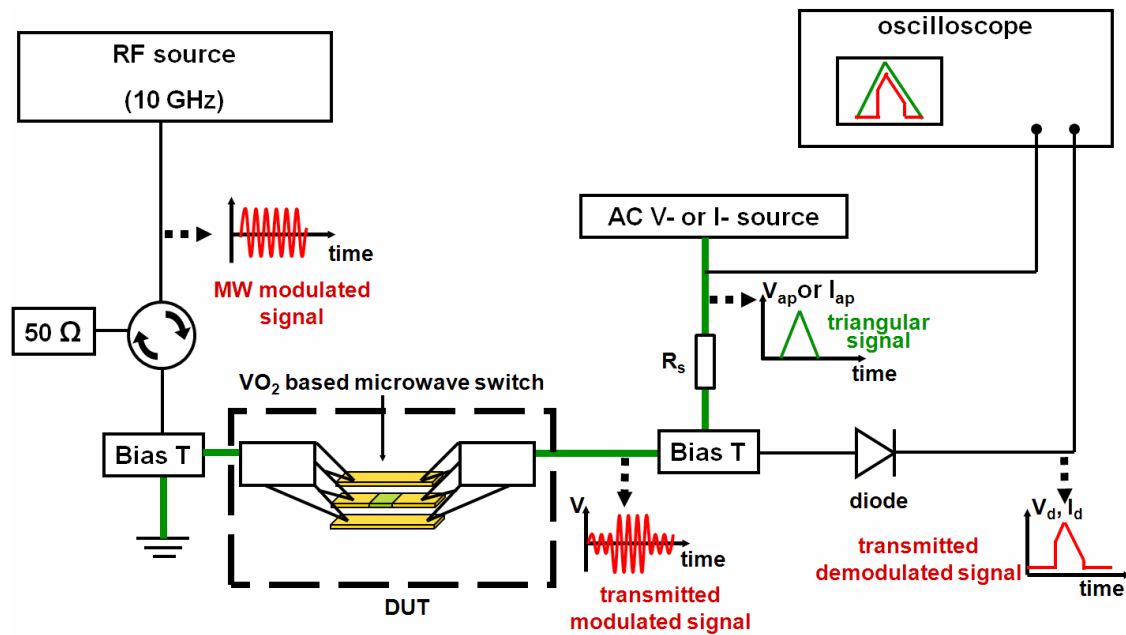
Figure 2(b) shows the dynamic behavior of a similar  $\text{VO}_2$ -based switch with a 20- $\mu\text{m}$ -long  $\text{VO}_2$  pattern, expressed as the amplitude variation of the transmitted MW signal ( $S_{21}$  parameter) when the MIT transition of the  $\text{VO}_2$  pattern is periodically triggered using an ac signal with triangular waveform, 100 V amplitude and 10 Hz frequency. In this case,



**Figure 2.** (a) Frequency dependence of the  $S_{21}$  transmission parameter for a CPW switch in the off state ( $\text{VO}_2$  SC) and in the on state ( $\text{VO}_2$  metal). The inset shows an optical microscopy image of the switch. (b) Periodic variation of the  $S_{21}$  parameter for a similar CPW switch (with a 20- $\mu\text{m}$ -long  $\text{VO}_2$  strip) cycled between the off and on states using a triangular waveform (100 V amplitude, 10 Hz frequency).

the  $S_{21}$  amplitude varies between the two extreme values that correspond to the semiconducting and metallic states of  $\text{VO}_2$  (dashed curves in figure 2(b)).

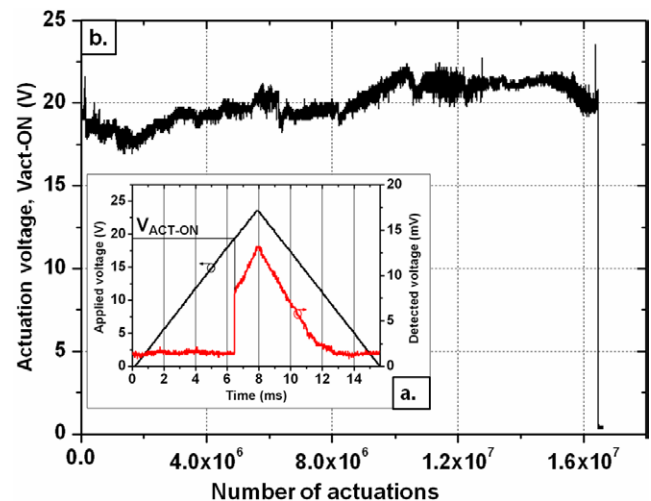
Devices having the characteristics shown in figure 2 were used for investigating the change in their properties during cycling activation of the  $\text{VO}_2$  pattern, in both V- and I-modes. They were integrated in the setup shown in figure 3. It consists of an MW source (Agilent E3633A) operating at 10 GHz—the frequency where the CPW switches have  $\sim 20$  dB difference in signal transmission between the on and off states, as observed in figures 2(a) and (b)—connected to the device under test (DUT) via a circulator and a 10 GHz band-pass filter. The  $\text{VO}_2$ -based device is introduced in a Desert Microwave probe station under dry  $\text{N}_2$  atmosphere. The MW signal traveling through the device is detected by a detector (Agilent, B474C) whose output is recorded on an oscilloscope. The MIT in  $\text{VO}_2$  is activated by applying a periodic low-frequency voltage or current signal to the two parts of the central signal line of the device, which are interrupted by the  $\text{VO}_2$  pattern (inset in figure 2(a)), using a pair of bias Ts. A 500- $\Omega$  resistance is included in the external circuit, in series with the  $\text{VO}_2$  switch. The activation signal is displayed and recorded on a second channel of the oscilloscope. A computer program automatically acquires the applied low-frequency signal and



**Figure 3.** Schematic of the setup used for assessing the lifetime of 10- $\mu\text{m}$ -long  $\text{VO}_2$  switches.

the detected MW signal transmitted through the device, allowing the recorded results to be presented in the form shown in figure 4(a) for a voltage-activated device.

**3.2.1 Voltage-controlled activation.** We first tested the performance of the device shown in the inset of figure 2(a) in the voltage activation mode. For activating the 10- $\mu\text{m}$ -long  $\text{VO}_2$  switch, we applied to the device, via the bias Ts, a triangular voltage waveform of 43 Hz frequency and 23.5 V amplitude. As indicated above, when  $\text{VO}_2$  is a semiconductor, the MW signal is considered in the off state: the 10 GHz MW signal cannot propagate and no voltage is recorded from the MW detector. For voltages above the threshold value  $V_s$  in figure 1 or  $V_{\text{ACT-ON}}$  in figure 4(a),  $\text{VO}_2$  becomes a metal, the MW signal propagates through the device, the switch is considered in the on state, and a signal is recorded from the MW detector. As shown in figure 4(a), the detected MW signal represented by the red curve allows recording the activation voltage of the device ( $V_{\text{ACT-ON}}$ ) as an indicator of the MIT transition. The second threshold voltage of the control signal,  $V_{\text{ACT-OFF}}$ , which converts  $\text{VO}_2$  back to a semiconductor and corresponds to the  $M^*$  point in figure 1(a), has not been recorded because its value was too low and comparable with the noise amplitude. Figure 4(b) shows a typical evolution of  $V_{\text{ACT-ON}}$  values versus the number of activation cycles.  $V_{\text{ACT-ON}}$  remains relatively stable for over 16.25 million activation cycles, with less than 20% variation, mainly because of the MW source instability, and then abruptly drops to zero indicating the device failure. Inspection of the device under an optical microscope revealed a clear degradation of the  $\text{VO}_2$  layer, manifested by a change in color and exfoliation, which was likely induced by heat accumulation during the cycling. Indeed, as suggested above, the large hysteresis of the V-mode trace in figure 1(a), and

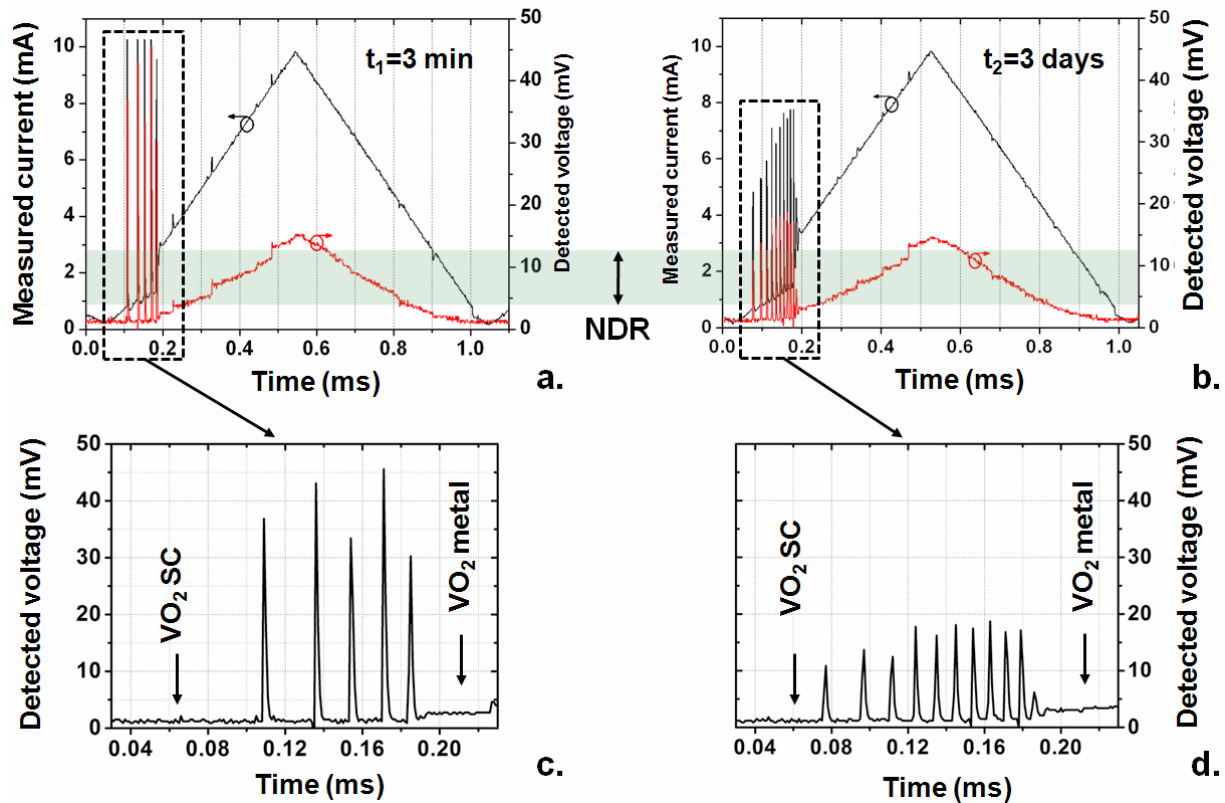


**Figure 4.** (a) Typical actuation cycle of a voltage-activated 10- $\mu\text{m}$ -long  $\text{VO}_2$  switch: the applied voltage signal (black curve) and the MW signal (red curve) detected with the setup of figure 3. (b) Evolution of  $V_{\text{ACT-ON}}$  with the number of actuation cycles.

correspondingly, the large hysteresis of the detector response in figure 4(a), suggest a thermally controlled MIT transition.

**3.2.2 Current-controlled activation.** A similar test was performed on an identical device activated in the I-mode. We used a current source (Keithley 6221) delivering a triangular waveform at 1 kHz frequency and 10 mA amplitude to the 10- $\mu\text{m}$ -long  $\text{VO}_2$  switch. The applied input current was displayed on the oscilloscope as the potential drop on a 500  $\Omega$  series resistance. Prior  $I$ - $V$  measurement in the I-mode showed that the threshold current  $I_{\text{ACT-ON}}$  ( $I_s$  in figure 1) triggering the MIT is  $\sim 1$  mA. Figures 5(a) and (b) show in a similar manner as figure 4(a) the evolution





**Figure 5.** Applied current (black curves, 1 kHz frequency) and MW voltage (red curves) detected from a 10- $\mu\text{m}$ -long  $\text{VO}_2$  device, (a) after 3 min of actuation and (b) after more than 250 million actuation cycles. (c) and (d) Magnification of the dashed rectangles in (a) and (b), respectively.

of the detected MW signal (red curves) together with the applied signal ('measured current on  $R_s$ ', black curves), at the beginning of the test (after 3 min, figure 5(a)) and after more than 250 million current-actuation cycles (figure 5(b)). The device sustained more than 260 million cycles with no visible degradation, and further testing was interrupted merely for technical reasons.

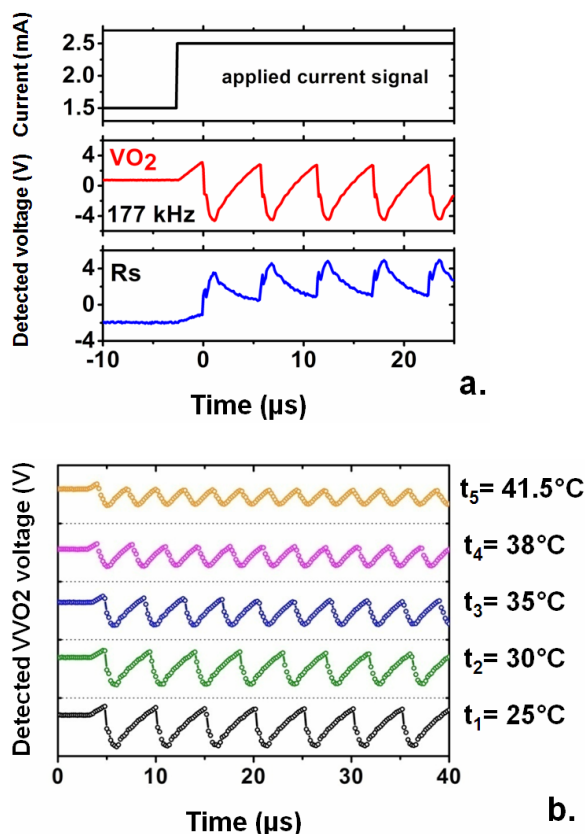
The reliability tests described above were conducted on three types of CPW device, having the  $\text{VO}_2$  pattern lengths between 10 and 20  $\mu\text{m}$ , in the voltage- and current-controlled regimes. Overall, the lifetime was at least 16 times longer when the  $\text{VO}_2$  devices were activated in the current mode than in the voltage mode. As shown in figures 1(a) and (b), the  $I$ - $V$  hysteresis is far more pronounced for the V-mode activation; the hysteresis width in this case is defined by the Joule heating when  $\text{VO}_2$  is in the metallic state (current values above 45 mA). The hysteresis was narrower for the I-mode activation because of less pronounced resistive heating, as  $\text{VO}_2$  becomes metallic at currents above 5 mA. In this case,  $\text{VO}_2$  experiences lower thermal stress and lower electric fields, which should reduce the risk of dielectric breakdown, explaining the lifetime enhancement of the current-activated devices.

An interesting feature visible in figures 5(a) and (b) is current-induced self-oscillations in  $\text{VO}_2$ . These oscillations are marked by dashed rectangles in figures 5(a) and (b) and are magnified in figures 5(c) and (d), respectively; they are related to the NDR region in the  $I$ - $V$  curves between

1 and 2.8 mA. Their occurrence depends on several external parameters, such as the amplitude of the activation signal and the value of the series resistance, and was previously observed only for voltage-controlled  $\text{VO}_2$  switches [4, 21]. Their onset was explained in terms of the percolative MIT in  $\text{VO}_2$ , namely, the coexistence of insulator-metallic domains seen as periodic construction-destruction of capacitive phases in the material [4, 20].

The evolution of the detected amplitude and frequency of these self-oscillations during the reliability test is the only sign of fatigue of the tested device. Indeed, as observed in figures 5(c) and (d), the amplitude of self-oscillations decreases from about 42 mV at the beginning of the test to about 17 mV after 250 million current-actuation cycles, while their average frequency increases from  $\sim 54$  to  $\sim 84$  kHz. These variations can be explained by the resistivity change of  $\text{VO}_2$  or by heat accumulation in the device owing to successive current injection. Indeed, this phenomenon was observed in experiments where we induced self-oscillations in a two-terminal  $\text{VO}_2$  switch of 5- $\mu\text{m}$  length and 20- $\mu\text{m}$  width by applying a square-shaped current signal (100 Hz, 2 mA amplitude) within the NDR region of the device. The applied current signals as well as the oscillations induced in the  $\text{VO}_2$  switch, represented as voltage detected across the series resistance, are plotted in figure 6(a).

The amplitude and frequency of these oscillations depend, among other parameters, on the temperature of the  $\text{VO}_2$  switch, as shown in figure 6(b) for the 5- $\mu\text{m}$ -long



**Figure 6.** (a) Typical self-oscillations induced by a square-shaped current waveform (100 Hz, 2 mA amplitude) in a 5- $\mu\text{m}$ -long, 20- $\mu\text{m}$ -wide two-terminal VO<sub>2</sub> switch; panel (b) shows the oscillations with temperature from  $t_1 = 25^\circ\text{C}$  to  $t_5 = 41.5^\circ\text{C}$ .

VO<sub>2</sub> switch for temperatures ranging from  $t_1 = 25^\circ\text{C}$  to  $t_5 = 41.5^\circ\text{C}$ . The amplitude of the oscillations decreases by 60% (from 6.6 to 2.6 V) with increasing temperature from  $t_1$  to  $t_5$ , whereas their frequency increases by 40% (from  $\sim 0.2$  to  $\sim 0.3$  MHz). The variation of the parameters of the self-oscillations appearing during the current-activation of VO<sub>2</sub> switches is thus a very fine indicator of the changes induced in the VO<sub>2</sub> material. This property can be further exploited for the use of these devices as temperature, pressure or gas sensors.

#### 4. Conclusions

We investigated the reliability and lifetime of the MIT in two-terminal VO<sub>2</sub> switches under multiple cycles of voltage- and current-induced activation. The lifetime was at least 16 times longer for the current-driven than the voltage-driven VO<sub>2</sub> devices; current-induced activation is less affected by thermal effects, resulting in a smaller degradation of the VO<sub>2</sub> films. This activation scheme is accompanied by electrical self-oscillations induced in the VO<sub>2</sub>-based switch, which are

a fine indicator of the modification of material properties. The presented results demonstrate the potential for the integration of VO<sub>2</sub> thin films in devices for advanced applications requiring a large number of stable and reproducible switching cycles.

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